

LABORATOIRE DE PHYSIQUE DES LASERS

Atelier d'électronique

cnrs

dépasser les frontières

Double Drivers IGBT sécurisés

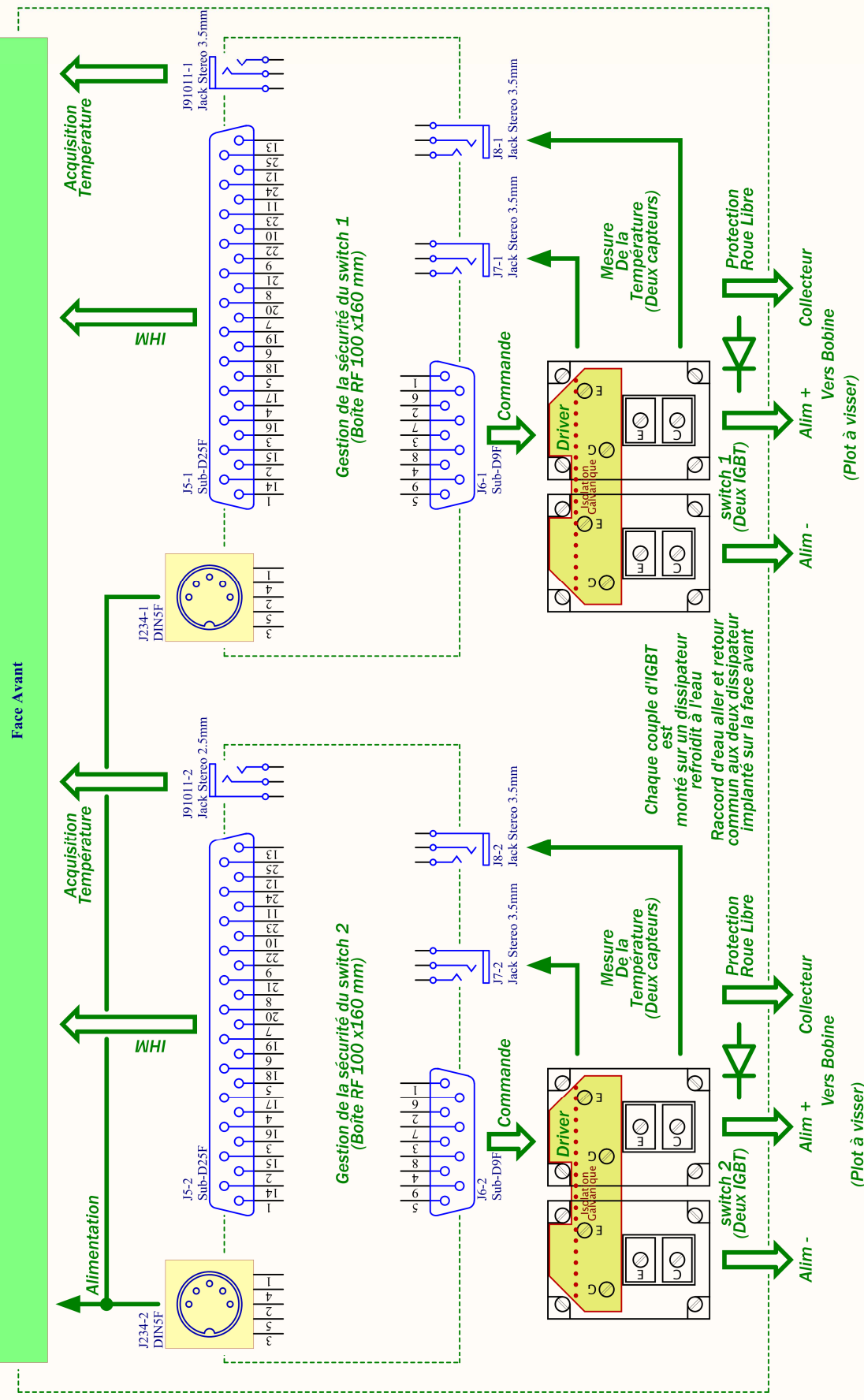
Description

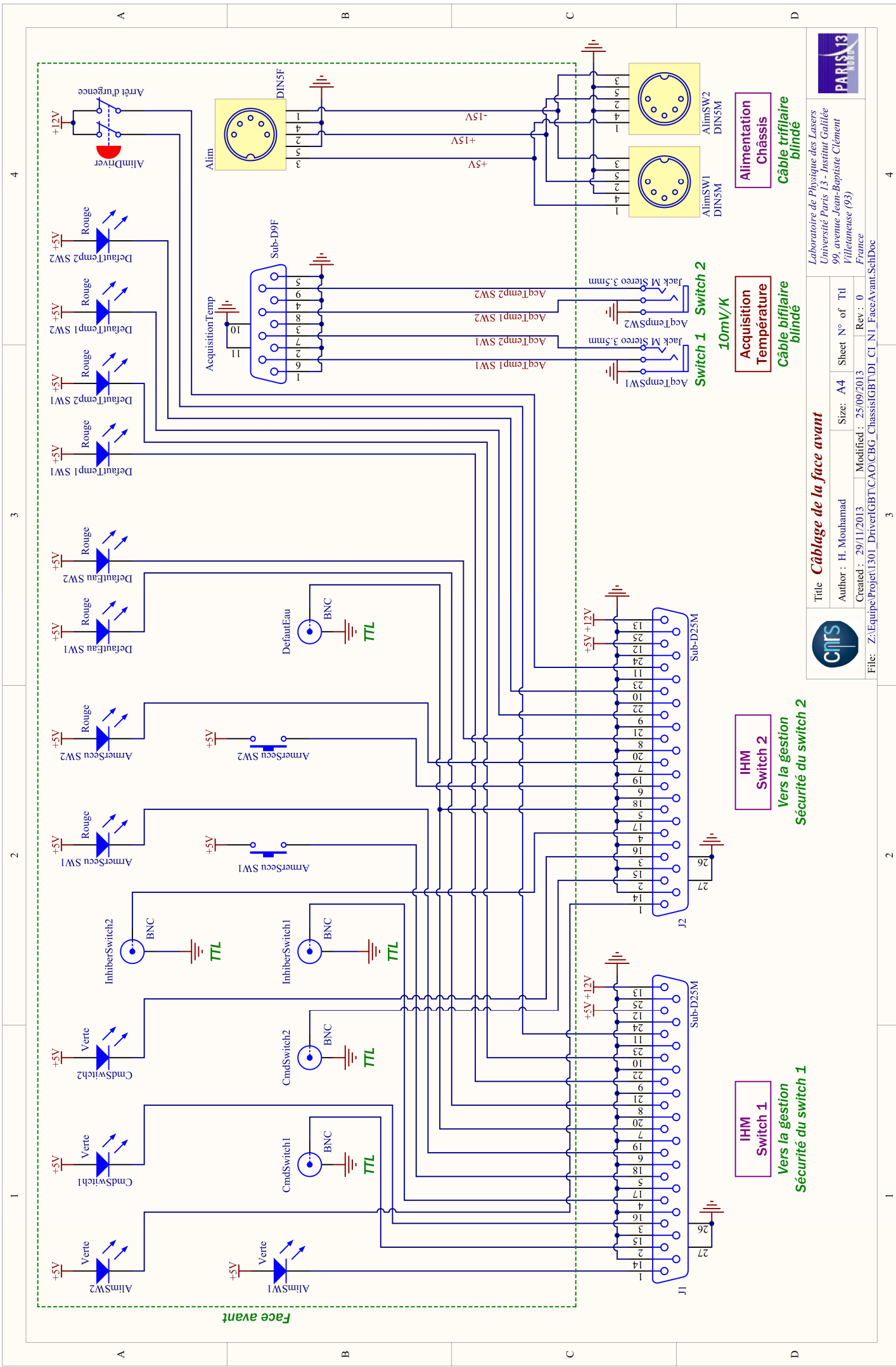
Réf. : DriverIGBT

Equipe : BEC

Auteur : Haniffe MOUHAMAD

FA
DL_CI_NI_FaceAvant.SchDoc





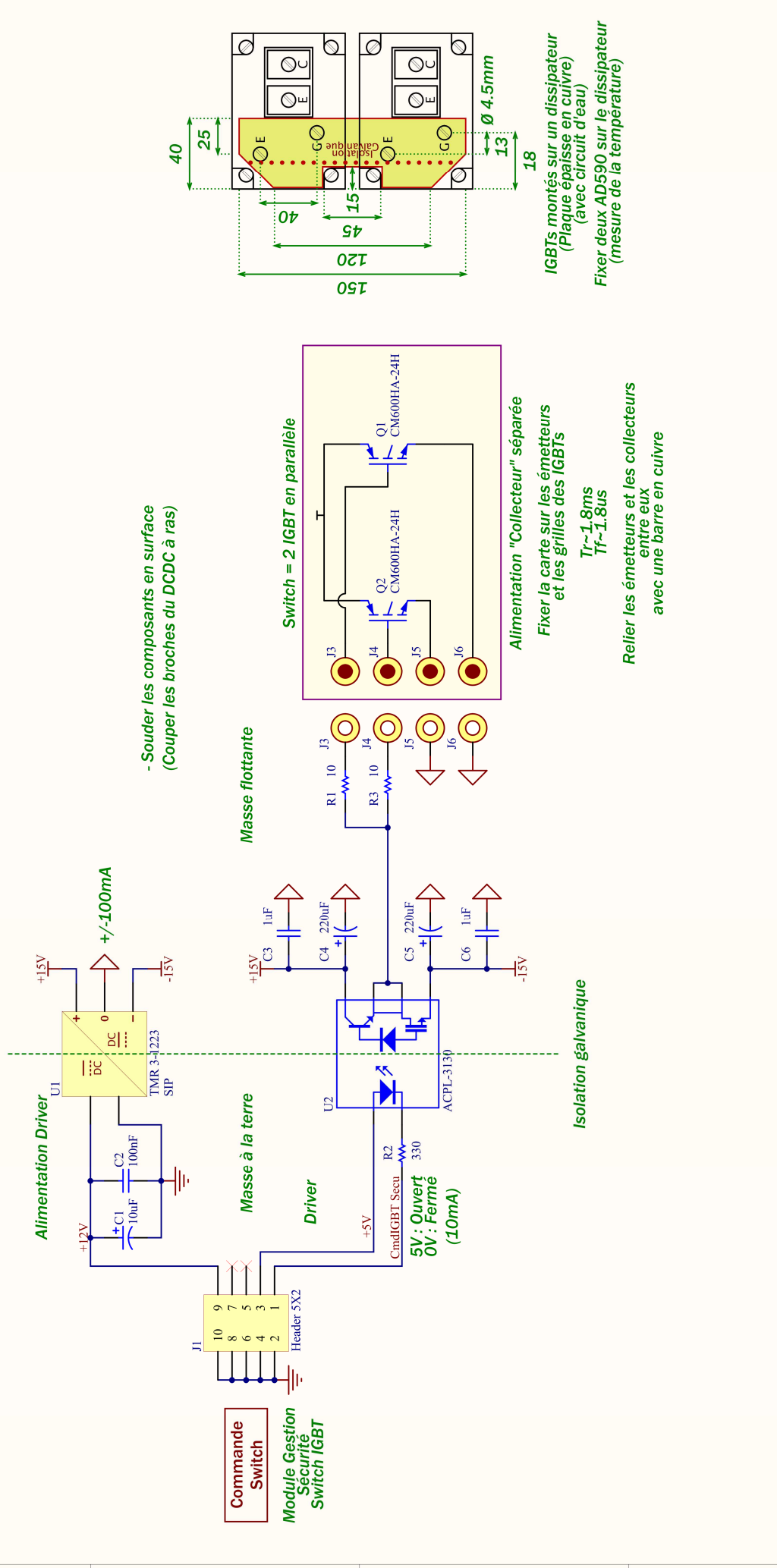
Face avant

IHM
Switch 1
Vers la gestion
Sécurité du switch 1

IHM
Switch 2
Vers la gestion
Sécurité du switch 2

Acquisition
Température
Câble bifilaire
blindé
10mV/K

Alimentation
Châssis
Câble trifilaire
blindé



- Solder les composants en surface
(Couper les broches du DCDC à ras)

Switch = 2 IGBT en parallèle

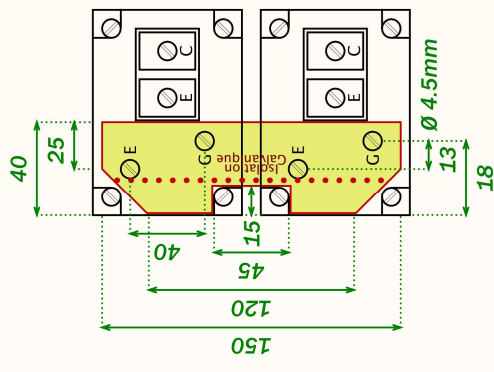
Alimentation "Collecteur" séparée

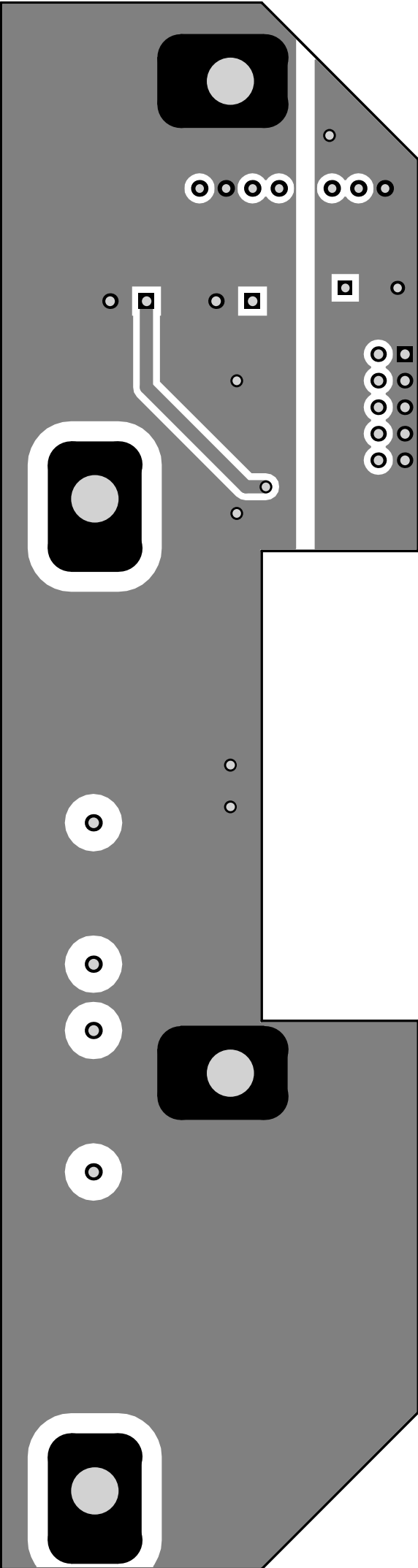
Fixer la carte sur les émetteurs et les grilles des IGBTs

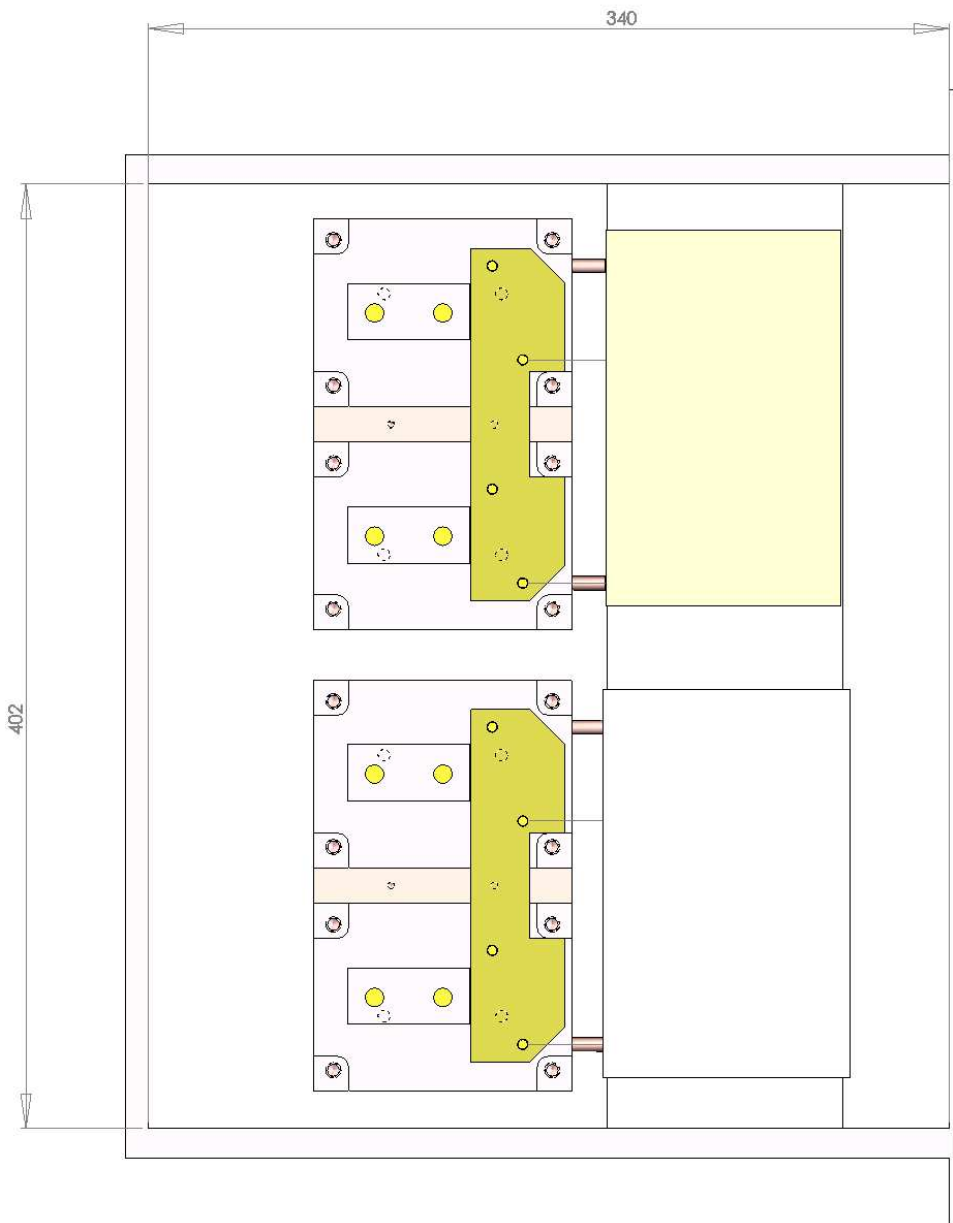
$T_r \sim 1.8ms$
 $T_f \sim 1.8\mu s$

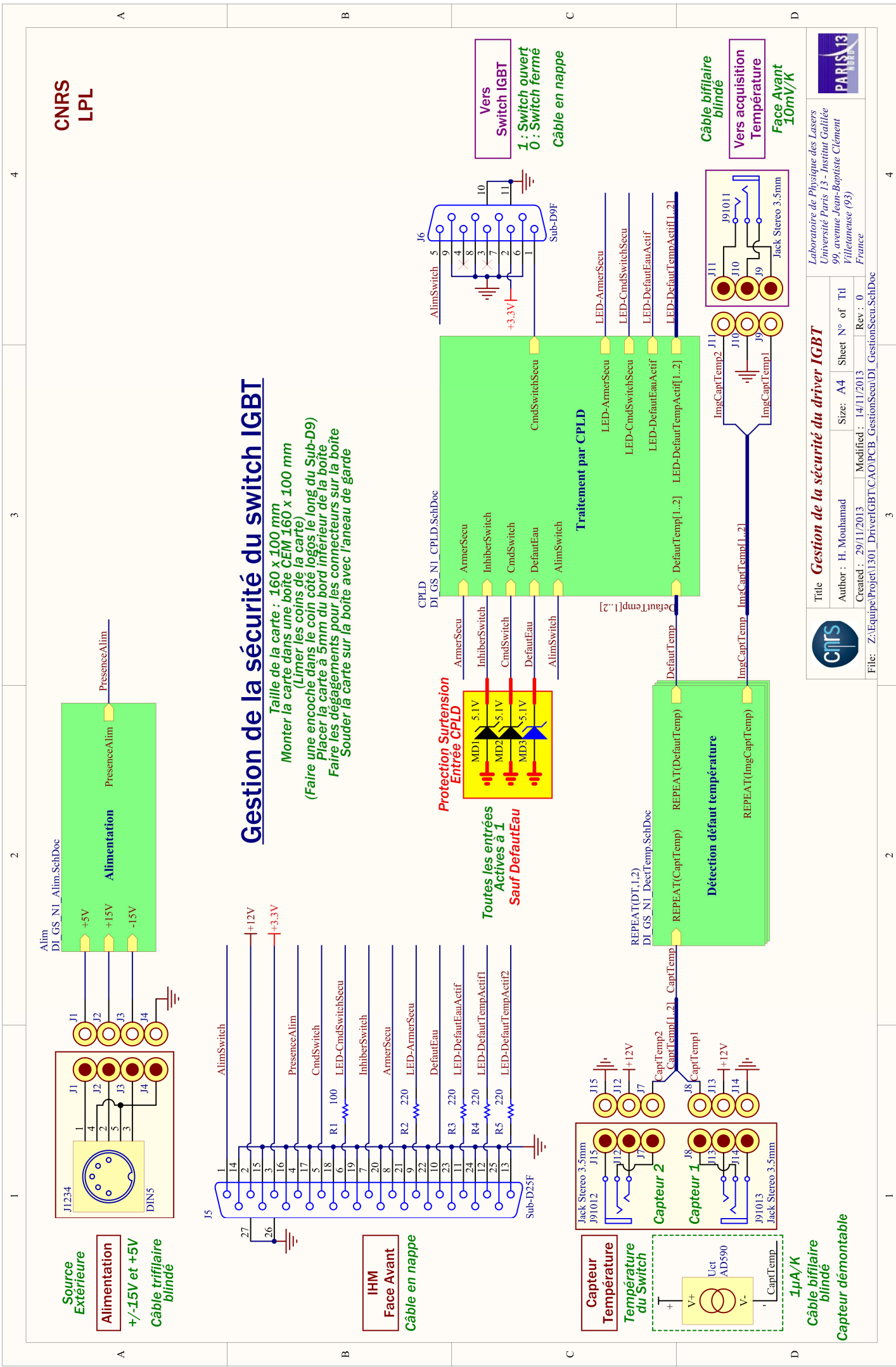
Relier les émetteurs et les collecteurs entre eux avec une barre en cuivre

IGBTs montés sur un dissipateur (Plaque épaisse en cuivre) (avec circuit d'eau)
Fixer deux AD590 sur le dissipateur (mesure de la température)









Source Exterieur
Alimentation
+/-15V et +5V
Câble trifilaire blindé

Gestion de la sécurité du switch IGBT

Taille de la carte : 160 x 100 mm
 Monter la carte dans une boîte CEM 160 x 100 mm
 (Limer les coins de la carte)
 (Faire une encoche dans le coin côté logis le long du Sub-D9)
 Placer la carte à 5mm du bord inférieur de la boîte
 Faire les dégagements pour les connecteurs sur la boîte
 Souder la carte sur la boîte avec l'anneau de garde

IHM
Face Avant
Câble en nappe

Toutes les entrées Actives à 1
Sauf DefautEau

Vers Switch IGBT
1 : Switch ouvert
0 : Switch fermé
Câble en nappe

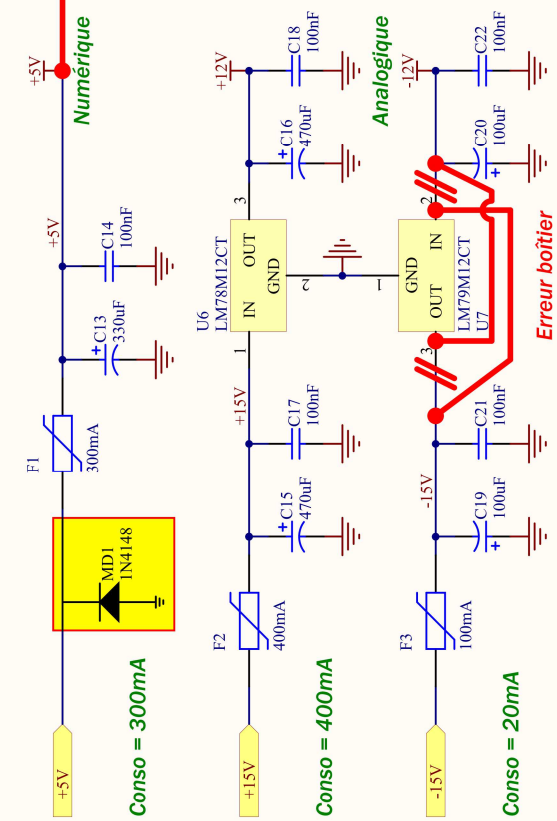
Capteur Temperature
Temperature du Switch
1µA/K
Câble bifilaire blindé
Capteur démontable

Câble bifilaire blindé
Vers acquisition Temperature
Face Avant
10mV/K

Title Gestion de la sécurité du driver IGBT	
Author : H. Mouhamad	Sheet N° of Ttl
Created : 29/11/2013	Size: A4
Modified : 14/11/2013	Revised : 0
File: Z:\Equipe\Projet\1301_Driver\GBT\CAO\PCB_GestionSecu\DI_GestionSecu.SchDoc	

Alimentation

Source
Externeure
Au châssis



Conso = 300mA

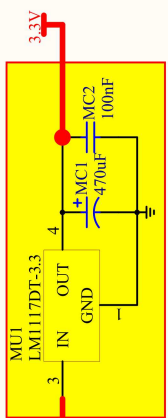
Conso = 400mA

Conso = 20mA

Arrivée
Alimentation

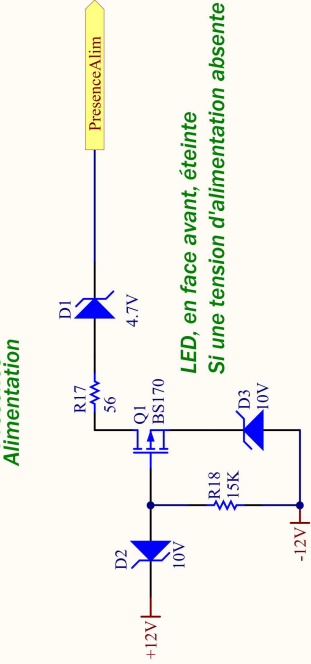
Analogique

Numérique



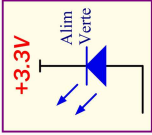
CPLD XC9536XL
doit être alimenté en +3.3V

Détection
Présence
Alimentation



LED, en face avant, éteinte
Si une tension d'alimentation absente

Indication
Présence
Alimentation
Face Avant



Title **Alimentation**

Author : H. Mouhamad

Size: A4

Sheet N° of Ttl

Created : 29/11/2013

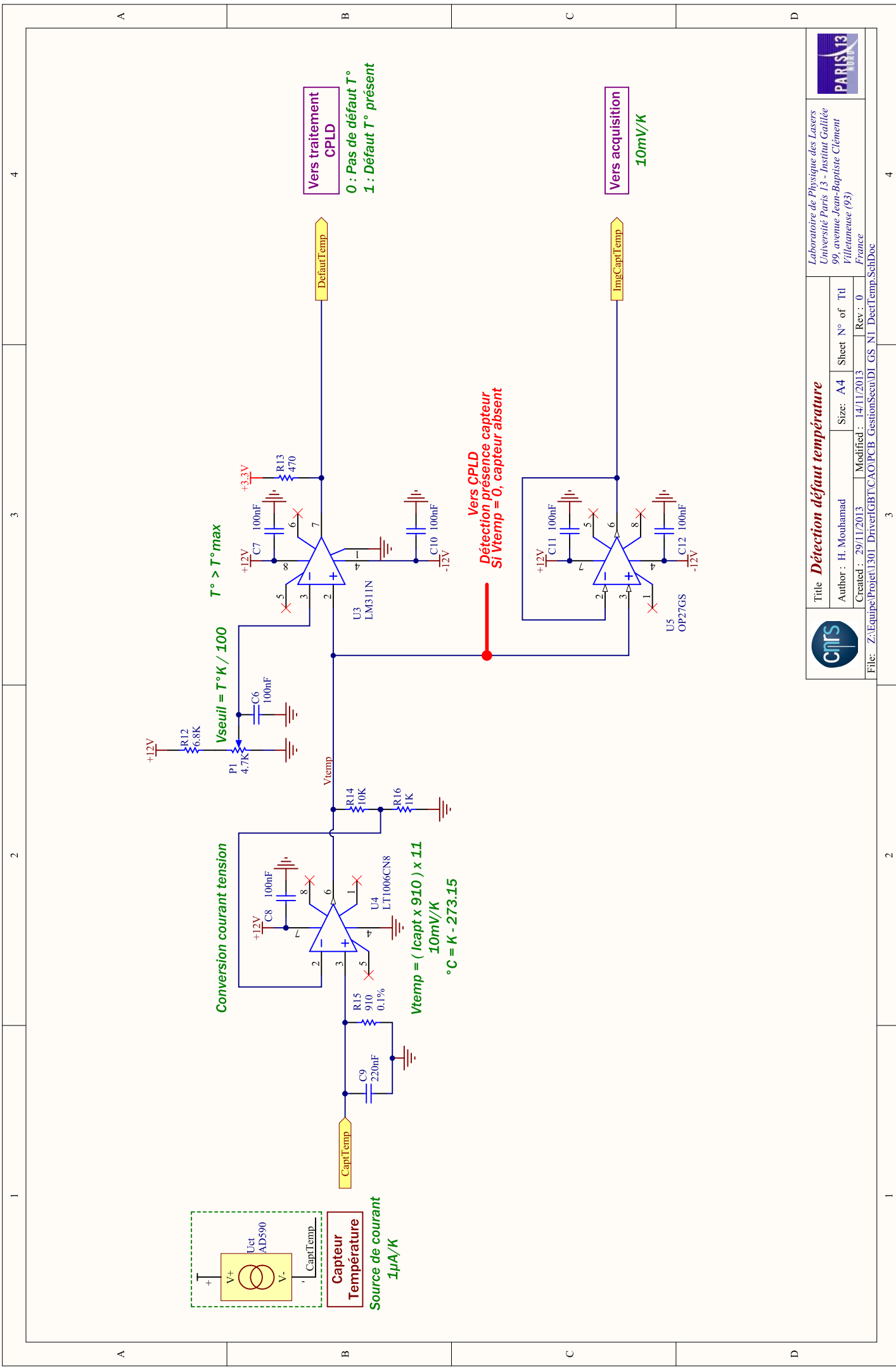
Modified : 14/11/2013

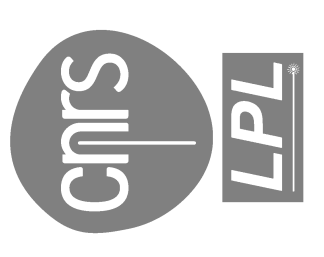
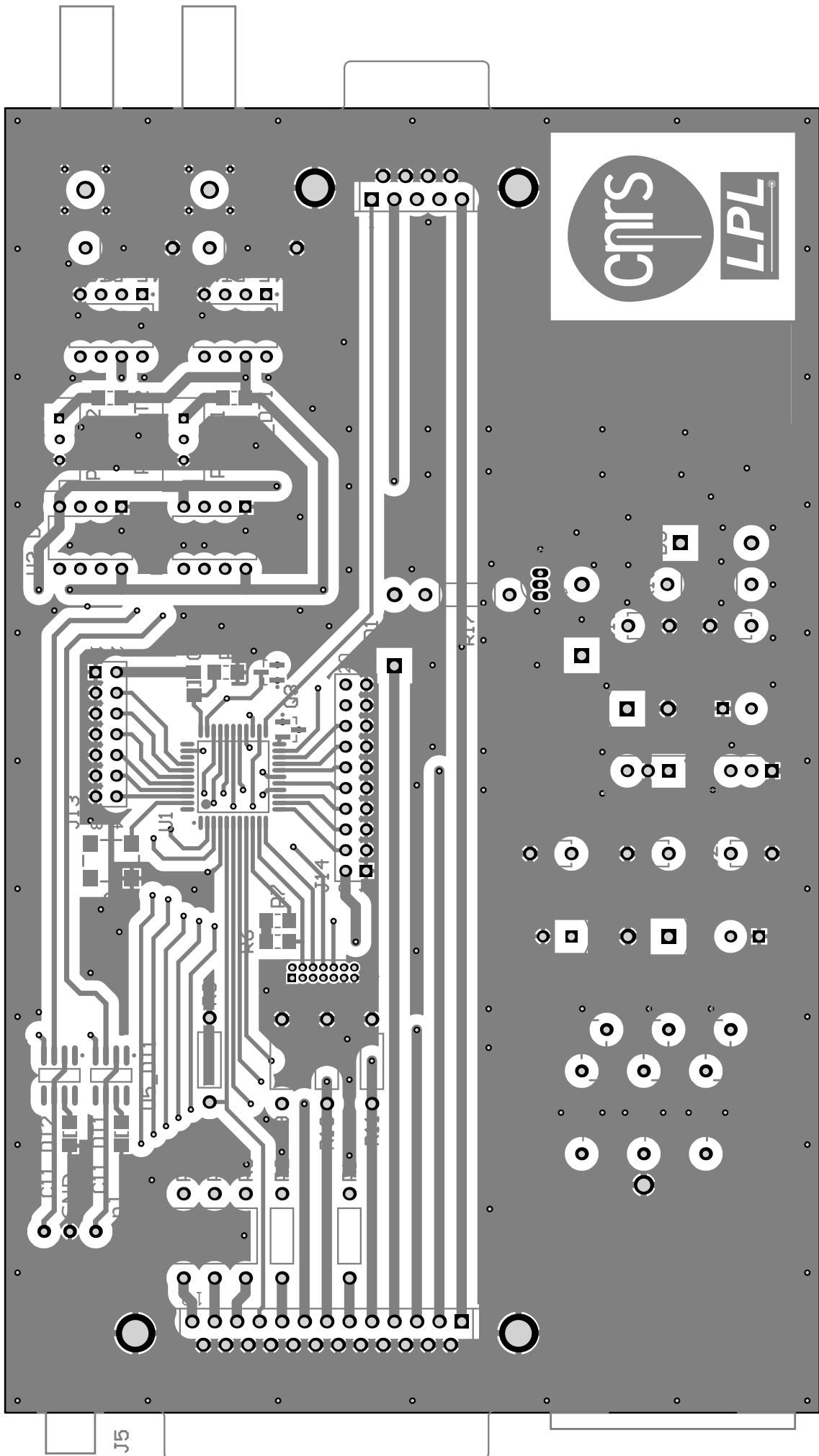
Rev : 0

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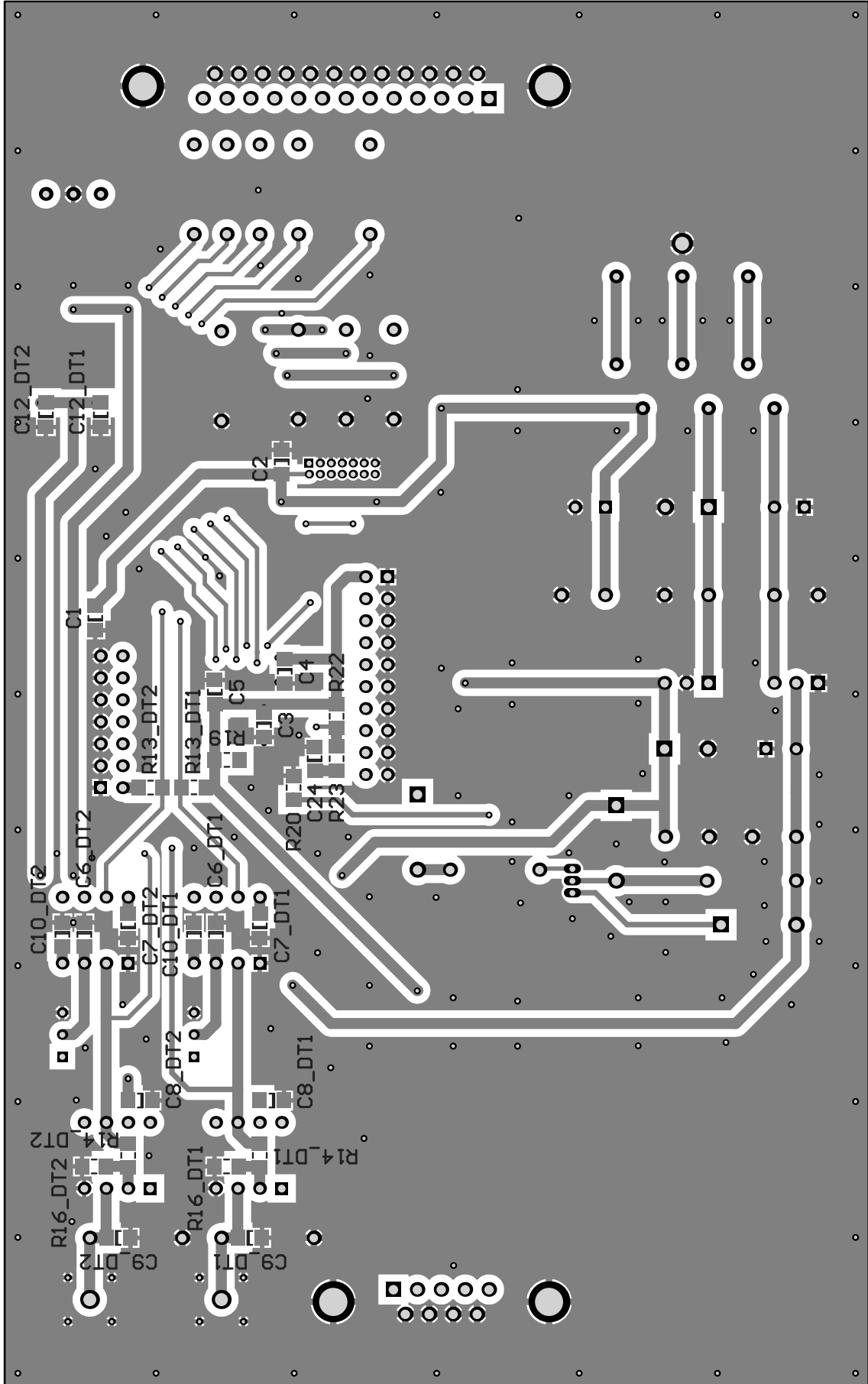
Laboratoire de Physique des Lasers
Université Paris 13 - Institut Galilée
99, avenue Jean-Baptiste Clément
Villetaneuse (93)
France

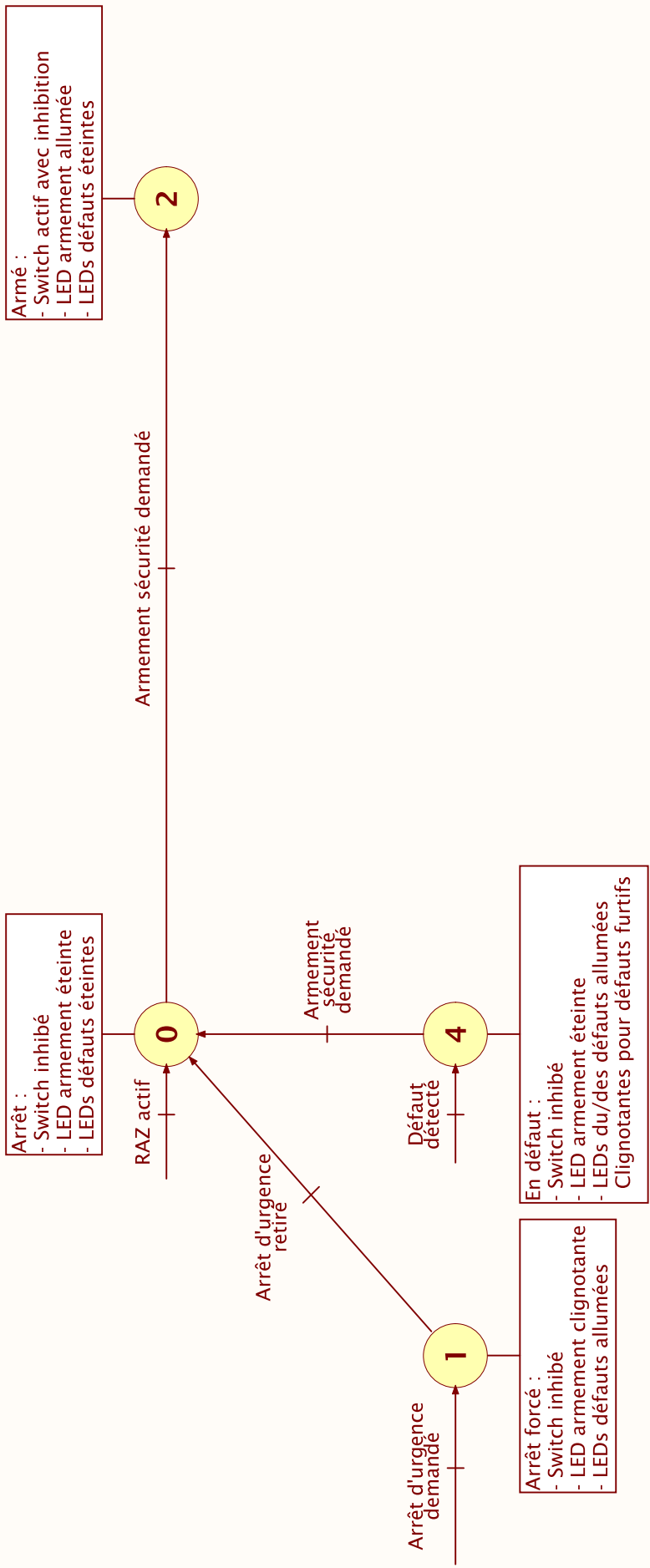






J5





Titre **Diagramme d'états du CPLD**

Author : H. Mouhammad	Size: A4	Sheet N° of T11
Created : 29/11/2013	Modified : 06/06/2013	Rev : 0

Laboratoire de Physique des Lasers
 Université Paris 13 - Institut Galilée
 99, avenue Jean-Baptiste Clément
 Villeneuve (93)
 France



File: Z:\Equipe\Projet\1301_Driver\GBT\CAO\SYNO_CPLD\DI DgEtat CPLD.SelDoc

```
-----
1  -- CNRS - Laboratoire de Physique des Lasers - Université Paris 13 Villetaneuse
2  -- Haniffe Mouhamad
3  --
4  --
5  -- Composant : XC9536
6  -- Projet : Driver_IGBT
7  --
8  -- GESTION DE LA SECURITE DE COMMANDE D'UN SWITCH A IGBT
9  --
10 -- Avril 2013
11 --
12 -----
13 library IEEE;
14 use IEEE.STD_LOGIC_1164.ALL;
15 use IEEE.std_logic_arith.all;
16 use IEEE.std_logic_unsigned.all;
17
18
19
20
21 -- ENTITE DE BASE
22 entity GestionSecu is
23     Port ( RAZ : in STD_LOGIC;
24           H : in STD_LOGIC;
25
26           EntreeOsc : out STD_LOGIC;
27           SortieOsc : in STD_LOGIC;
28           EtatOsc : out STD_LOGIC;
29
30           ArretUrgence : in STD_LOGIC;
31           InhiberSwitch : in STD_LOGIC;
32           ArmerSecu : in STD_LOGIC;
33           LED_ArmerSecu : out STD_LOGIC;
34
35           -- Ordre des défauts
36           -- Temp 1 + Temp 2 + Eau
37           -- '1' : Défaut présent
38           Defaults : in STD_LOGIC_VECTOR (2 downto 0);
39           LED_Defaults : out STD_LOGIC_VECTOR (2 downto 0);
40           -- CaptTemp 1 + CaptTemp 2
41           -- '0' : Capteur absent
42           PresenceCaptTemp : in STD_LOGIC_VECTOR (1 downto 0);
```

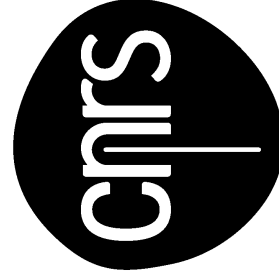
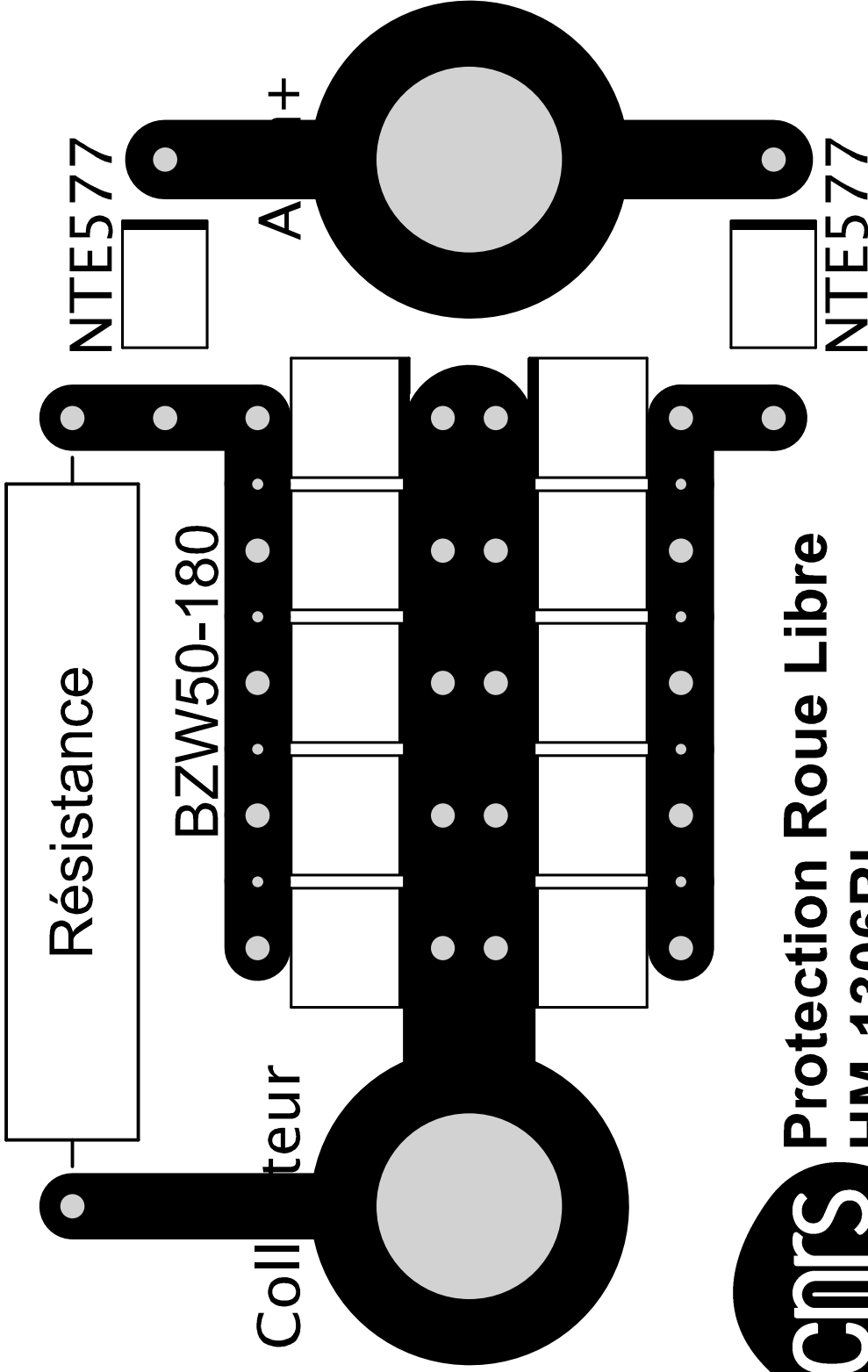
```
43 CmdSwitch : in STD_LOGIC;
44 CmdSwitchSecu : out STD_LOGIC;
45 LED_CmdSwitchSecu : out STD_LOGIC;
46 end GestionSecu;
47
48
49
50
51
52
53 -- ARCHITECTURE
54 architecture Behavioral of GestionSecu is
55
56 -- DECLARATION
57
58 -- Type énuméré
59 type ETATS is (ARRET, ARRET_FORCE, ARMEE, EN_DEFAULT);
60
61 -- Signaux internes
62 signal Etat : ETATS;
63 signal MemoDefaults : STD_LOGIC_VECTOR (2 downto 0);
64 signal RegDefaults : STD_LOGIC_VECTOR (2 downto 0);
65 signal Clignotante : STD_LOGIC;
66
67 -- Constantes
68 Constant LED_ETEINTE : STD_LOGIC := '1';
69 Constant LED_ALLUMEE : STD_LOGIC := '0';
70 Constant SWITCH_OUVERT : STD_LOGIC := '1';
71 Constant ACTIF : STD_LOGIC := '1';
72 Constant INACTIF : STD_LOGIC := '0';
73
74 begin
75
76
77 -- MACHINE D'ETATS
78
79 Process (H, RAZ)
80 begin
81
82
83 -- Initialisation par remise à zéro (Priorité 1)
84 if RAZ = ACTIF then
```



```
85 Etat <= ARRET;
86 CmdSwitchSecu <= SWITCH_OUVERT;
87 LED_CmdSwitchSecu <= LED_ETEINTE;
88 LED_ArmerSecu <= LED_ETEINTE;
89 LED_Defaults <= LED_ETEINTE & LED_ETEINTE & LED_ETEINTE;
90 MemoDefaults <= "111";
91
92
93
94 -- Arrêt d'urgence (Priorité 2)
95 elsif ArretUrgence = ACTIF then
96
97 Etat <= ARRET_FORCE;
98 CmdSwitchSecu <= SWITCH_OUVERT;
99 LED_CmdSwitchSecu <= LED_ETEINTE;
100 LED_ArmerSecu <= Clignotante;
101 LED_Defaults <= LED_ALLUMEE & LED_ALLUMEE & LED_ALLUMEE;
102
103
104 -- Défauts (Priorité 3)
105 -- '0' : Défaut présent
106 -- Allumage de la LED défaut concernée
107 elsif RegDefaults < "111" then
108
109 Etat <= EN_DEFAULT;
110
111 -- Mise à jour des défauts
112 if RegDefaults(0)='0' then MemoDefaults(0) <= '0'; end if;
113 if RegDefaults(1)='0' then MemoDefaults(1) <= '0'; end if;
114 if RegDefaults(2)='0' then MemoDefaults(2) <= '0'; end if;
115
116 CmdSwitchSecu <= SWITCH_OUVERT;
117 LED_CmdSwitchSecu <= LED_ETEINTE;
118 LED_ArmerSecu <= LED_ETEINTE;
119 LED_Defaults <= ( ( Clignotante & Clignotante & Clignotante ) or MemoDefaults ) and RegDefaults;
120
121
122 -- ETAT DE LA MACHINE
123 elsif H'Event and H='1' then
124 case Etat is
125
126
```

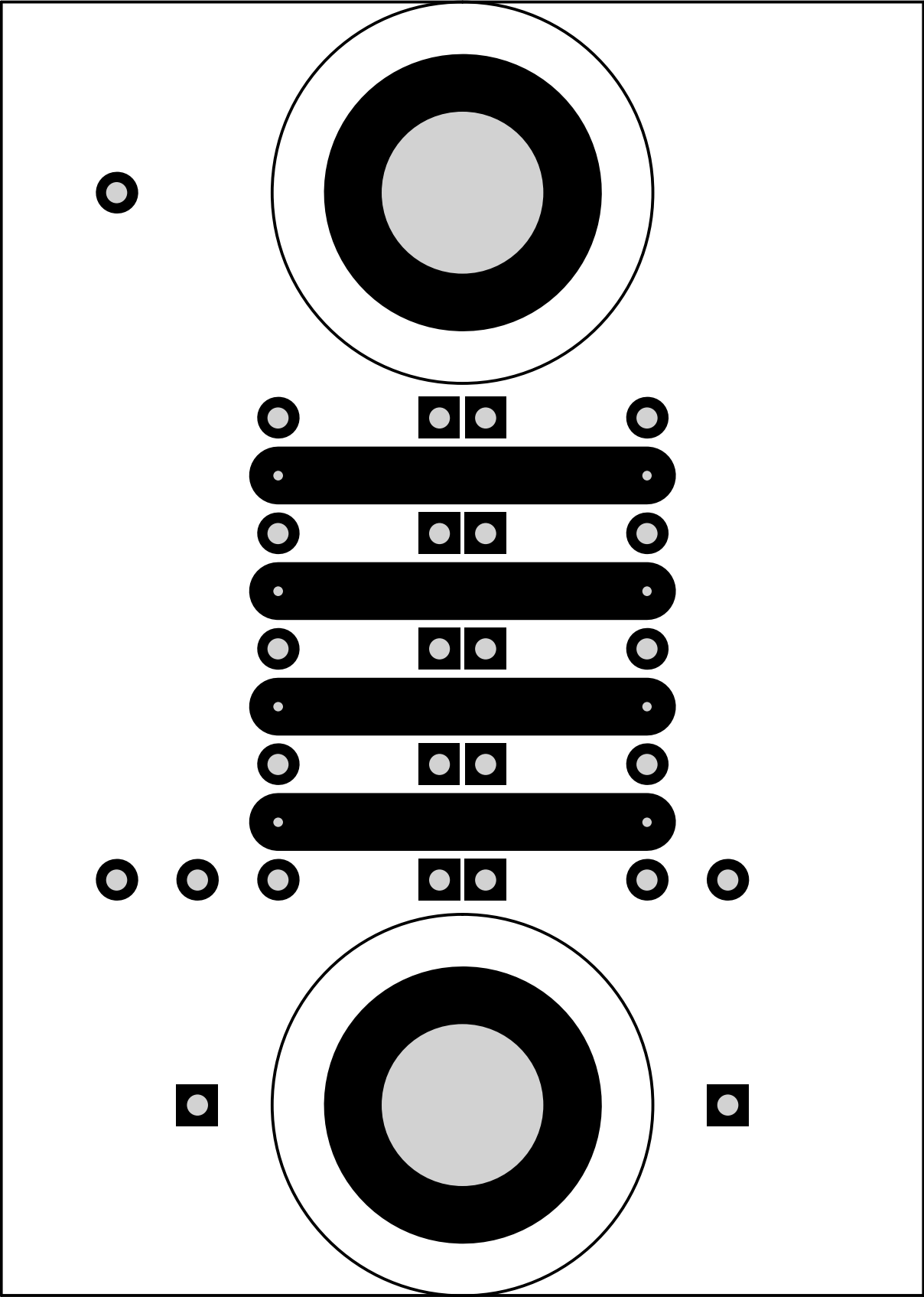
```
127 when ARRET_FORCE =>
128   if ArretUrgence = ACTIF then Etat <= ARRET_FORCE;
129   else Etat <= ARRET;
130   end if;
131
132
133 when ARRET =>
134   CmdSwitchSecu <= SWITCH_OUVERT;
135   LED_CmdSwitchSecu <= LED_ETEINTE;
136   LED_ArmerSecu <= LED_ETEINTE;
137   LED_Defaults <= LED_ETEINTE & LED_ETEINTE & LED_ETEINTE;
138
139   if ArmerSecu = ACTIF then Etat <= ARMEE;
140   else Etat <= ARRET;
141   end if;
142
143
144 when ARMEE =>
145   Etat <= ARMEE;
146   CmdSwitchSecu <= not ( CmdSwitch and not InhiberSwitch );
147   LED_CmdSwitchSecu <= not ( CmdSwitch and not InhiberSwitch );
148   LED_ArmerSecu <= LED_ALLUMEE or ( InhiberSwitch and Clignotante );
149   LED_Defaults <= LED_ETEINTE & LED_ETEINTE & LED_ETEINTE;
150
151
152 when EN_DEFAULT => -- LED défaut(s) concernées(s) clignotantes
153   CmdSwitchSecu <= SWITCH_OUVERT;
154   LED_CmdSwitchSecu <= LED_ETEINTE;
155   LED_ArmerSecu <= LED_ETEINTE;
156   LED_Defaults <= ( Clignotante & Clignotante & Clignotante ) or MemoDefaults;
157
158   if RegDefaults = "111" and ArmerSecu = ACTIF then Etat <= ARRET;
159   MemoDefaults <= "111";
160   else Etat <= EN_DEFAULT;
161   end if;
162
163
164 when others => null;
165   end case;
166   end if;
167   end process;
168
```

```
169
170 -- Clignotement à 410ms
171 Clignotante <= SortieOsc;
172
173
174 -- Oscillateur
175 -- Période 410 ms
176 EntreeOsc <= not SortieOsc;
177 EtatOsc <= SortieOsc;
178
179
180 -- Registre des défauts
181 -- Temp 1 + Temp 2 + Eau
182 -- CaptTemp 1 + CaptTemp 2
183 -- '0' : Défaut présent
184 RegDefaults(2) <= Defaults(2); -- Si Défaut actif 0, ne pas mettre "not"
185 RegDefaults(1) <= not Defaults(1) and PresenceCaptTemp(1);
186 RegDefaults(0) <= not Defaults(0) and PresenceCaptTemp(0);
187
188
189 end Behavioral;
190
191
```

Protection Roue Libre
HM-1306RL







Driver sécurisé Double switches IGBT



SW1 Alim

SW2 Alim

SW1

Fermé

Cmd

SW2

Fermé

Cmd

Sécurité

SW1

Armer

Inhiber

SW2

Armer

Inhiber

Défaut Eau

SW1

Détection

SW2

Défaut Température

SW1

DT1

DT2

DT1

DT2

Lecture Temp

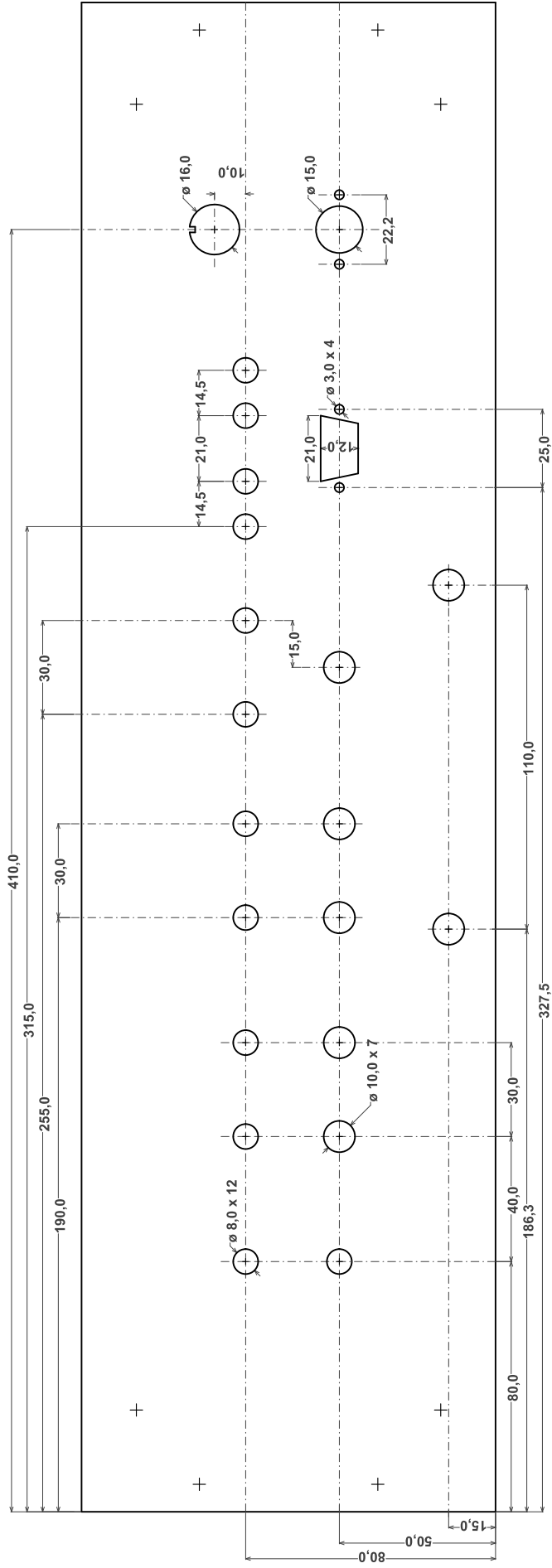
Arrêt d'urgence

Alimentation

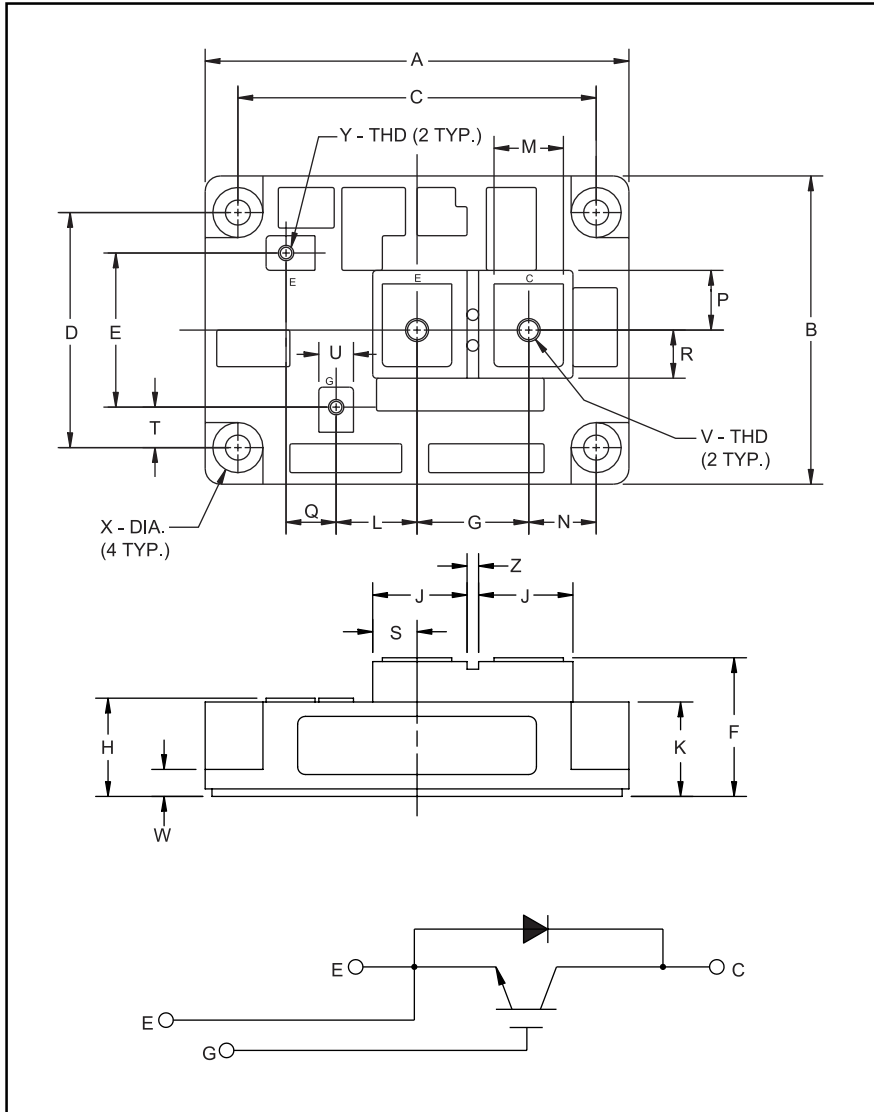


Dissipation par eau





Single IGBTMOD™ H-Series Module 600 Amperes/1200 Volts



Outline Drawing and Circuit Diagram

Dimensions	Inches	Millimeters
A	4.33	110.0
B	3.15	80.0
C	3.66±0.008	93.0±0.25
D	2.44±0.008	62.0±0.25
E	1.57	40.0
F	1.42 Max.	36.0 Max.
G	1.14	29.0
H	1.00 Max.	25.5 Max.
J	0.96	25.0
K	0.94	24.5
L	0.83	21.0
M	0.71	18.0

Dimensions	Inches	Millimeters
N	0.69	17.5
P	0.61	15.5
Q	0.51	13.0
R	0.49	12.5
S	0.45	11.5
T	0.43	11.0
U	0.35	9.0
V	M8 Metric	M8
W	0.28	7.0
X	0.256 Dia.	Dia. 6.50
Y	M4 Metric	M4
Z	0.12	3.04



Description:

Powerex IGBTMOD™ Modules are designed for use in switching applications. Each module consists of one IGBT Transistor in a single configuration with a reverse-connected super-fast recovery free-wheel diode. All components and interconnects are isolated from the heat sinking baseplate, offering simplified system assembly and thermal management.

Features:

- Low Drive Power
- Low $V_{CE(sat)}$
- Discrete Super-Fast Recovery (135ns) Free-Wheel Diode
- High Frequency Operation (20-25kHz)
- Isolated Baseplate for Easy Heat Sinking

Applications:

- AC Motor Control
- Motion/Servo Control
- UPS
- Welding Power Supplies
- Laser Power Supplies

Ordering Information:

Example: Select the complete part module number you desire from the table below -i.e. CM600HA-24H is a 1200V (V_{CES}), 600 Ampere Single IGBTMOD™ Power Module.

Type	Current Rating Amperes	V_{CES} Volts (x 50)
CM	600	24



Powerex, Inc., 200 Hillis Street, Youngwood, Pennsylvania 15697-1800 (724) 925-7272

CM600HA-24H
Single IGBTMOD™ H-Series Module
 600 Amperes/1200 Volts

Absolute Maximum Ratings, $T_j = 25\text{ }^\circ\text{C}$ unless otherwise specified

Ratings	Symbol	CM600HA-24H	Units
Junction Temperature	T_j	-40 to 150	$^\circ\text{C}$
Storage Temperature	T_{stg}	-40 to 125	$^\circ\text{C}$
Collector-Emitter Voltage (G-E SHORT)	V_{CES}	1200	Volts
Gate-Emitter Voltage	V_{GES}	± 20	Volts
Collector Current	I_C	600	Amperes
Peak Collector Current	I_{CM}	1200*	Amperes
Diode Forward Current	I_F	600	Amperes
Diode Forward Surge Current	I_{FM}	1200*	Amperes
Power Dissipation	P_d	4100	Watts
Max. Mounting Torque M8 Terminal Screws	-	95	in-lb
Max. Mounting Torque M6 Mounting Screws	-	26	in-lb
Module Weight (Typical)	-	560	Grams
V Isolation	V_{RMS}	2500	Volts

* Pulse width and repetition rate should be such that device junction temperature does not exceed the device rating.

Static Electrical Characteristics, $T_j = 25\text{ }^\circ\text{C}$ unless otherwise specified

Characteristics	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Collector-Cutoff Current	I_{CES}	$V_{CE} = V_{CES}, V_{GE} = 0V$	-	-	2.0	mA
Gate Leakage Current	I_{GES}	$V_{GE} = V_{GES}, V_{CE} = 0V$	-	-	0.5	μA
Gate-Emitter Threshold Voltage	$V_{GE(th)}$	$I_C = 60\text{mA}, V_{CE} = 10V$	4.5	6.0	7.5	Volts
Collector-Emitter Saturation Voltage	$V_{CE(sat)}$	$I_C = 600\text{A}, V_{GE} = 15V$	-	2.5	3.4**	Volts
		$I_C = 600\text{A}, V_{GE} = 15V, T_j = 150^\circ\text{C}$	-	2.25	-	Volts
Total Gate Charge	Q_G	$V_{CC} = 600V, I_C = 600\text{A}, V_{GS} = 15V$	-	3000	-	nC
Diode Forward Voltage	V_{FM}	$I_E = 600\text{A}, V_{GS} = 0V$	-	-	3.5	Volts

** Pulse width and repetition rate should be such that device junction temperature rise is negligible.

Dynamic Electrical Characteristics, $T_j = 25\text{ }^\circ\text{C}$ unless otherwise specified

Characteristics	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Input Capacitance	C_{ies}		-	-	120	nF
Output Capacitance	C_{oes}	$V_{GE} = 0V, V_{CE} = 10V, f = 1\text{MHz}$	-	-	42	nF
Reverse Transfer Capacitance	C_{res}		-	-	24	nF
Resistive	Turn-on Delay Time	$V_{CC} = 600V, I_C = 600\text{A},$				300
	Rise Time					
Switching	Turn-off Delay Time	$V_{GE1} = V_{GE2} = 15V, R_G = 2.1\Omega$				450
	Fall Time					
Diode Reverse Recovery Time	t_{rr}	$I_E = 600\text{A}, di_E/dt = -1200\text{A}/\mu\text{s}$	-	-	250	ns
Diode Reverse Recovery Charge	Q_{rr}	$I_E = 600\text{A}, di_E/dt = -1200\text{A}/\mu\text{s}$	-	4.46	-	μC

Thermal and Mechanical Characteristics, $T_j = 25\text{ }^\circ\text{C}$ unless otherwise specified

Characteristics	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Thermal Resistance, Junction to Case	$R_{th(j-c)}$	Per IGBT	-	-	0.035	$^\circ\text{C}/\text{W}$
Thermal Resistance, Junction to Case	$R_{th(j-c)}$	Per FWDi	-	-	0.06	$^\circ\text{C}/\text{W}$
Contact Thermal Resistance	$R_{th(c-f)}$	Per Module, Thermal Grease Applied	-	-	0.035	$^\circ\text{C}/\text{W}$

Features

- 5 ns pin-to-pin logic delays
- System frequency up to 178 MHz
- 36 macrocells with 800 usable gates
- Available in small footprint packages
 - 44-pin PLCC (34 user I/O pins)
 - 44-pin VQFP (34 user I/O pins)
 - 48-pin CSP (36 user I/O pins)
 - 64-pin VQFP (36 user I/O pins)
 - Pb-free available for all packages
- Optimized for high-performance 3.3V systems
 - Low power operation
 - 5V tolerant I/O pins accept 5 V, 3.3V, and 2.5V signals
 - 3.3V or 2.5V output capability
 - Advanced 0.35 micron feature size CMOS Fast FLASH™ technology
- Advanced system features
 - In-system programmable
 - Superior pin-locking and routability with Fast CONNECT™ II switch matrix
 - Extra wide 54-input Function Blocks
 - Up to 90 product-terms per macrocell with individual product-term allocation
 - Local clock inversion with three global and one product-term clocks
 - Individual output enable per output pin
 - Input hysteresis on all user and boundary-scan pin inputs
 - Bus-hold circuitry on all user pin inputs
 - Full IEEE Standard 1149.1 boundary-scan (JTAG)
- Fast concurrent programming
- Slew rate control on individual outputs
- Enhanced data security features
- Excellent quality and reliability
 - Endurance exceeding 10,000 program/erase cycles
 - 20 year data retention
 - ESD protection exceeding 2,000V
- Pin-compatible with 5V-core XC9536 device in the 44-pin PLCC package and the 48-pin CSP package

WARNING: Programming temperature range of
 $T_A = 0^\circ \text{C}$ to $+70^\circ \text{C}$

Description

The XC9536XL is a 3.3V CPLD targeted for high-performance, low-voltage applications in leading-edge communications and computing systems. It is comprised of two

54V18 Function Blocks, providing 800 usable gates with propagation delays of 5 ns. See [Figure 2](#) for architecture overview.

Power Estimation

Power dissipation in CPLDs can vary substantially depending on the system frequency, design application and output loading. To help reduce power dissipation, each macrocell in a XC9500XL device may be configured for low-power mode (from the default high-performance mode). In addition, unused product-terms and macrocells are automatically deactivated by the software to further conserve power. For a general estimate of I_{CC} , the following equation may be used:

$$I_{CC}(\text{mA}) = MC_{HS}(0.175 \cdot PT_{HS} + 0.345) + MC_{LP}(0.052 \cdot PT_{LP} + 0.272) + 0.04 \cdot MC_{TOG}(MC_{HS} + MC_{LP}) \cdot f$$

where:

MC_{HS} = # macrocells in high-speed configuration

PT_{HS} = average number of high-speed product terms per macrocell

MC_{LP} = # macrocells in low power configuration

PT_{LP} = average number of low power product terms per macrocell

f = maximum clock frequency

MC_{TOG} = average % of flip-flops toggling per clock (~12%)

This calculation was derived from laboratory measurements of an XC9500XL part filled with 16-bit counters and allowing a single output (the LSB) to be enabled. The actual I_{CC} value varies with the design application and should be verified during normal system operation. [Figure 1](#) shows the above estimation in a graphical form. For a more detailed discussion of power consumption in this device, see Xilinx

application note [XAPP114, "Understanding XC9500XL CPLD Power."](#)

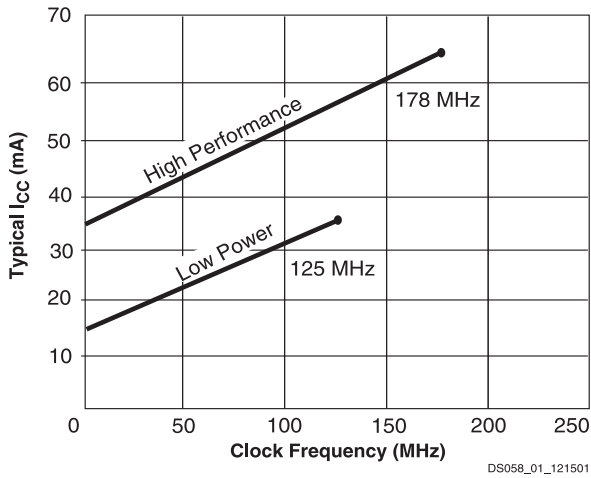
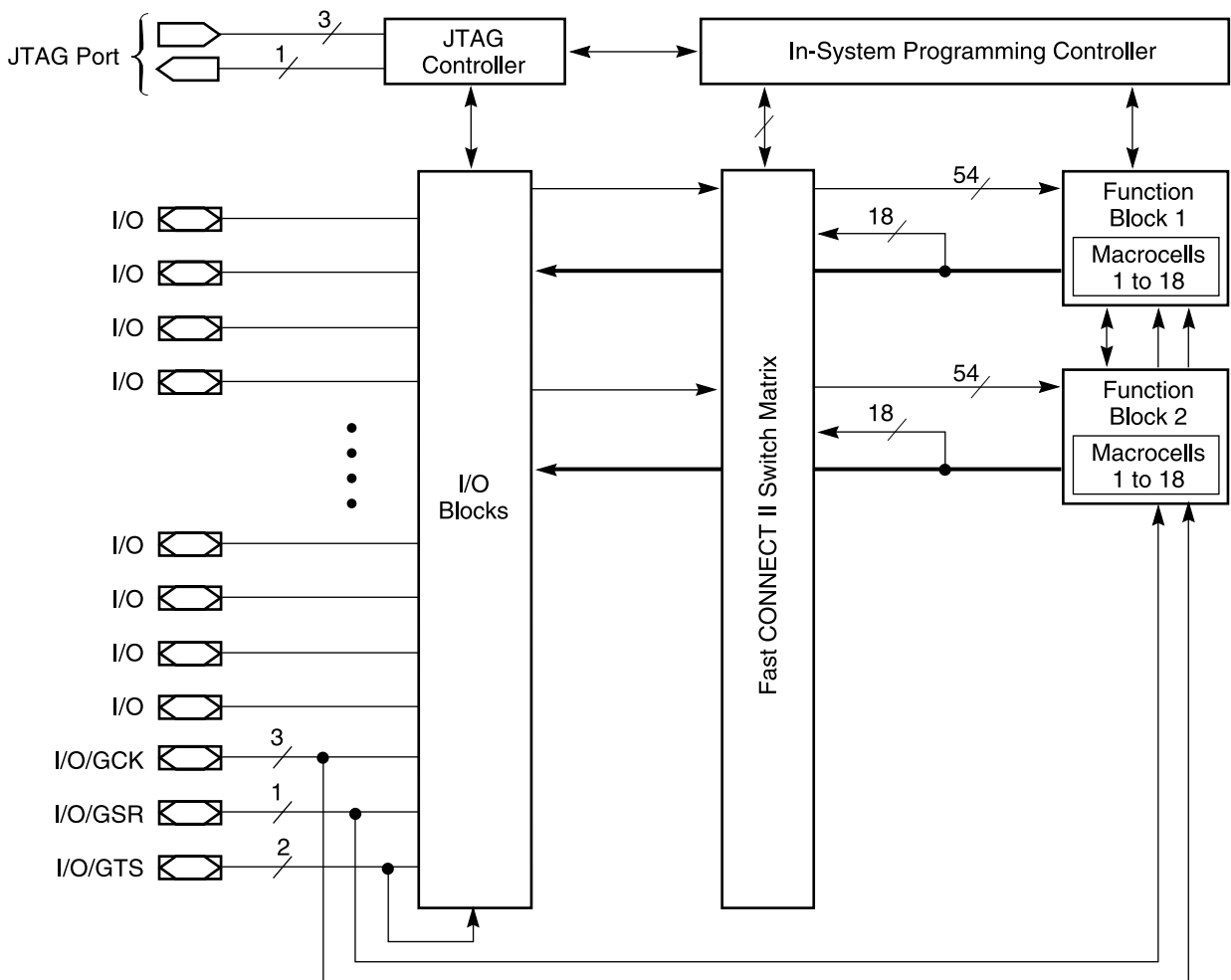


Figure 1: Typical I_{CC} vs. Frequency for XC9536XL



DS058_02_081500

Figure 2: XC9536XL Architecture
Function Block outputs (indicated by the bold line) drive the I/O Blocks directly.

Absolute Maximum Ratings⁽²⁾

Symbol	Description	Value	Units
V_{CC}	Supply voltage relative to GND	-0.5 to 4.0	V
V_{IN}	Input voltage relative to GND ⁽¹⁾	-0.5 to 5.5	V
V_{TS}	Voltage applied to 3-state output ⁽¹⁾	-0.5 to 5.5	V
T_{STG}	Storage temperature (ambient) ⁽³⁾	-65 to +150	°C
T_J	Junction temperature	+150	°C

Notes:

- Maximum DC undershoot below GND must be limited to either 0.5V or 10 mA, whichever is easier to achieve. During transitions, the device pins may undershoot to -2.0 V or overshoot to +7.0V, provided this over- or undershoot lasts less than 10 ns and with the forcing current being limited to 200 mA. External I/O voltage may not exceed V_{CCINT} by 4.0V.
- Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those listed under Operating Conditions is not implied. Exposure to Absolute Maximum Ratings conditions for extended periods of time may affect device reliability.
- For soldering guidelines and thermal considerations, see the [Device Packaging](#) information on the Xilinx website. For Pb-free packages, see [XAPP427](#).

Recommended Operation Conditions

Symbol	Parameter	Min	Max	Units	
V_{CCINT}	Supply voltage for internal logic and input buffers	Commercial $T_A = 0^\circ\text{C}$ to 70°C	3.0	3.6	V
		Industrial $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$	3.0	3.6	V
V_{CCIO}	Supply voltage for output drivers for 3.3V operation	3.0	3.6	V	
	Supply voltage for output drivers for 2.5V operation	2.3	2.7	V	
V_{IL}	Low-level input voltage	0	0.80	V	
V_{IH}	High-level input voltage	2.0	5.5	V	
V_O	Output voltage	0	V_{CCIO}	V	

Quality and Reliability Characteristics

Symbol	Parameter	Min	Max	Units
T_{DR}	Data Retention	20	-	Years
N_{PE}	Program/Erase Cycles (Endurance)	10,000	-	Cycles
V_{ESD}	Electrostatic Discharge (ESD)	2,000	-	Volts

DC Characteristic Over Recommended Operating Conditions

Symbol	Parameter	Test Conditions	Min	Max	Units
V_{OH}	Output high voltage for 3.3V outputs	$I_{OH} = -4.0$ mA	2.4	-	V
	Output high voltage for 2.5V outputs	$I_{OH} = -500$ μA	90% V_{CCIO}	-	V
V_{OL}	Output low voltage for 3.3V outputs	$I_{OL} = 8.0$ mA	-	0.4	V
	Output low voltage for 2.5V outputs	$I_{OL} = 500$ μA	-	0.4	V
I_{IL}	Input leakage current	$V_{CC} = \text{Max}; V_{IN} = \text{GND or } V_{CC}$	-	± 10	μA
I_{IH}	I/O high-Z leakage current	$V_{CC} = \text{Max}; V_{IN} = \text{GND or } V_{CC}$	-	± 10	μA
I_{IH}	I/O high-Z leakage current	$V_{CC} = \text{Max}; V_{CCIO} = \text{Max}; V_{IN} = \text{GND or } 3.6\text{V}$	-	± 10	μA
		$V_{CC} \text{ Min} < V_{IN} < 5.5\text{V}$	-	± 50	μA
C_{IN}	I/O capacitance	$V_{IN} = \text{GND}; f = 1.0$ MHz	-	10	pF
I_{CC}	Operating supply current (low power mode, active)	$V_{IN} = \text{GND}, \text{ No load}; f = 1.0$ MHz	10 (Typical)		mA

XC9536XL I/O Pins⁽²⁾

Function Block	Macro-cell	PC44	VQ44	CS48	VQ64	BScan Order	Function Block	Macro-cell	PC44	VQ44	CS48	VQ64	BScan Order
1	1	2	40	D6	9	105	2	1	1	39	D7	8	51
1	2	3	41	C7	10	102	2	2	44	38	E5	7	48
1	3	5 ⁽¹⁾	43 ⁽¹⁾	B7 ⁽¹⁾	15 ⁽¹⁾	99	2	3	42 ⁽¹⁾	36 ⁽¹⁾	E6 ⁽¹⁾	5 ⁽¹⁾	45
1	4	4	42	C6	11	96	2	4	43	37	E7	6	42
1	5	6 ⁽¹⁾	44 ⁽¹⁾	B6 ⁽¹⁾	16 ⁽¹⁾	93	2	5	40 ⁽¹⁾	34 ⁽¹⁾	F6 ⁽¹⁾	2 ⁽¹⁾	39
1	6	8	2	A6	19	90	2	6	39 ⁽¹⁾	33 ⁽¹⁾	G7 ⁽¹⁾	64 ⁽¹⁾	36
1	7	7 ⁽¹⁾	1 ⁽¹⁾	A7 ⁽¹⁾	17 ⁽¹⁾	87	2	7	38	32	G6	63	33
1	8	9	3	C5	20	84	2	8	37	31	F5	62	30
1	9	11	5	B5	22	81	2	9	36	30	G5	61	27
1	10	12	6	A4	24	78	2	10	35	29	F4	60	24
1	11	13	7	B4	25	75	2	11	34	28	G4	57	21
1	12	14	8	A3	27	72	2	12	33	27	E3	56	18
1	13	18	12	B2	33	69	2	13	29	23	F2	50	15
1	14	19	13	B1	35	66	2	14	28	22	G1	48	12
1	15	20	14	C2	36	63	2	15	27	21	F1	45	9
1	16	22	16	C3	38	60	2	16	26	20	E2	44	6
1	17	24	18	D2	42	57	2	17	25	19	E1	43	3
1	18	-	-	D3	39	54	2	18	-	-	E4	49	0

Notes:

1. Global control pin.
2. The pin-outs are the same for Pb-free versions of packages.

XC9536XL Global, JTAG and Power Pins⁽¹⁾

Pin Type	PC44	VQ44	CS48	VQ64
I/O/GCK1	5	43	B7	15
I/O/GCK2	6	44	B6	16
I/O/GCK3	7	1	A7	17
I/O/GTS1	42	36	E6	5
I/O/GTS2	40	34	F6	2
I/O/GSR	39	33	G7	64
TCK	17	11	A1	30
TDI	15	9	B3	28
TDO	30	24	G2	53
TMS	16	10	A2	29
V _{CCINT} 3.3V	21, 41	15, 35	C1, F7	3, 37
V _{CCIO} 2.5V/3.3V	32	26	G3	55
GND	10, 23, 31	4, 17, 25	A5, D1, F3	21, 41, 54
No Connects	-	-	C4, D4	1, 4, 12, 13, 14, 18, 23, 26, 31, 32, 34, 40, 46, 47, 51, 52, 58, 59

Notes:

1. The pin-outs are the same for Pb-free versions of packages.

FEATURES

Linear current output: 1 $\mu\text{A}/\text{K}$
Wide temperature range: -55°C to $+150^\circ\text{C}$
Probe-compatible ceramic sensor package
2-terminal device: voltage in/current out
Laser trimmed to $\pm 0.5^\circ\text{C}$ calibration accuracy (AD590M)
Excellent linearity: $\pm 0.3^\circ\text{C}$ over full range (AD590M)
Wide power supply range: 4 V to 30 V
Sensor isolation from case
Available in 2-lead FLATPACK, 4-lead LFCSP, 3-pin TO-52, 8-lead SOIC, and die form

GENERAL DESCRIPTION

The AD590 is a 2-terminal integrated circuit temperature transducer that produces an output current proportional to absolute temperature. For supply voltages between 4 V and 30 V, the device acts as a high impedance, constant current regulator passing 1 $\mu\text{A}/\text{K}$. Laser trimming of the chip's thin-film resistors is used to calibrate the device to 298.2 μA output at 298.2 K (25°C).

The AD590 should be used in any temperature-sensing application below 150°C in which conventional electrical temperature sensors are currently employed. The inherent low cost of a monolithic integrated circuit combined with the elimination of support circuitry makes the AD590 an attractive alternative for many temperature measurement situations. Linearization circuitry, precision voltage amplifiers, resistance measuring circuitry, and cold junction compensation are not needed in applying the AD590.

In addition to temperature measurement, applications include temperature compensation or correction of discrete components, biasing proportional to absolute temperature, flow rate measurement, level detection of fluids and anemometry. The AD590 is available in die form, making it suitable for hybrid circuits and fast temperature measurements in protected environments.

The AD590 is particularly useful in remote sensing applications. The device is insensitive to voltage drops over long lines due to its high impedance current output. Any well-insulated twisted pair is sufficient for operation at hundreds of feet from the receiving circuitry. The output characteristics also make the AD590 easy to multiplex: the current can be switched by a CMOS multiplexer, or the supply voltage can be switched by a logic gate output.

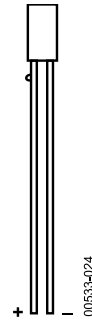
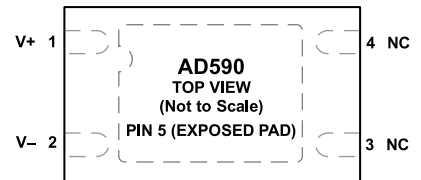
PIN CONFIGURATIONS


Figure 1. 2-Lead FLATPACK



NOTES
 1. NC = NO CONNECT. THE NC PIN IS NOT BONDED TO THE DIE INTERNALLY.
 2. TO ENSURE CORRECT OPERATION, THE EXPOSED PAD (EP) SHOULD BE LEFT FLOATING.

Figure 2. 4-Lead LFCSP

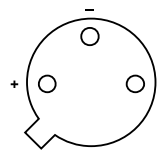


Figure 3. 3-Pin TO-52

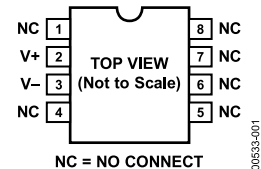


Figure 4. 8-Lead SOIC

PRODUCT HIGHLIGHTS

1. The AD590 is a calibrated, 2-terminal temperature sensor requiring only a dc voltage supply (4 V to 30 V). Costly transmitters, filters, lead wire compensation, and linearization circuits are all unnecessary in applying the device.
2. State-of-the-art laser trimming at the wafer level in conjunction with extensive final testing ensures that AD590 units are easily interchangeable.
3. Superior interface rejection occurs because the output is a current rather than a voltage. In addition, power requirements are low (1.5 mW @ 5 V @ 25°C). These features make the AD590 easy to apply as a remote sensor.
4. The high output impedance ($>10\text{ M}\Omega$) provides excellent rejection of supply voltage drift. For instance, changing the power supply from 5 V to 10 V results in only a 1 μA maximum current change, or 1°C equivalent error.
5. The AD590 is electrically durable: it withstands a forward voltage of up to 44 V and a reverse voltage of 20 V. Therefore, supply irregularities or pin reversal does not damage the device.

Rev. G

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GENERAL APPLICATIONS

Figure 17 shows a typical use of the AD590 in a remote temperature sensing application. The AD590 is used as a thermometer circuit that measures temperature from -55°C to $+150^{\circ}\text{C}$, with an output voltage of $1\text{ mV}/^{\circ}\text{K}$. Because the AD590 measures absolute temperature (its nominal output is $1\text{ mA}/\text{K}$), the output must be offset by 273.2 mA to read out in degrees Celsius.

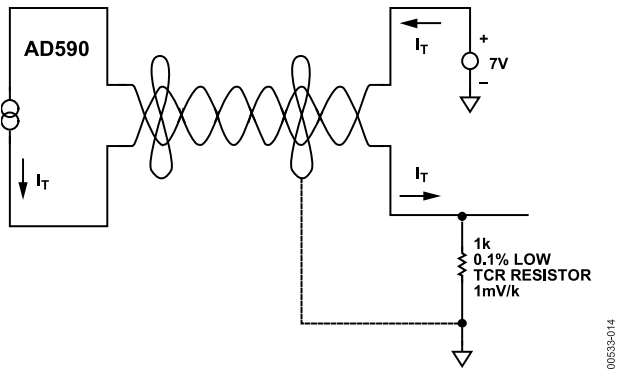


Figure 17. Variable Scale Display

Connecting several AD590 units in series, as shown in Figure 18, allows the minimum of all the sensed temperatures to be indicated. In contrast, using the sensors in parallel yields the average of the sensed temperatures.

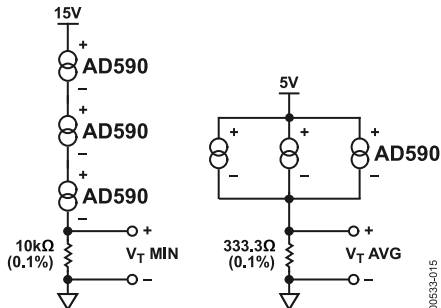


Figure 18. Series and Parallel Connection

The circuit in Figure 19 demonstrates one method by which differential temperature measurements can be made. R_1 and R_2 can be used to trim the output of the op amp to indicate a desired temperature difference. For example, the inherent offset between the two devices can be trimmed in. If V_+ and V_- are radically different, then the difference in internal dissipation causes a differential internal temperature rise. This effect can be used to measure the ambient thermal resistance seen by the sensors in applications such as fluid-level detectors or anemometry.

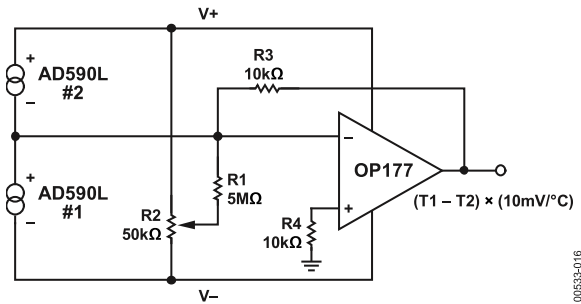


Figure 19. Differential Measurements

Figure 20 is an example of a cold junction compensation circuit for a Type J thermocouple using the AD590 to monitor the reference junction temperature. This circuit replaces an ice-bath as the thermocouple reference for ambient temperatures between 15°C and 35°C . The circuit is calibrated by adjusting R_T for a proper meter reading with the measuring junction at a known reference temperature and the circuit near 25°C . Using components with the TCs as specified in Figure 20, compensation accuracy is within $\pm 0.5^{\circ}\text{C}$ for circuit temperatures between 15°C and 35°C . Other thermocouple types can be accommodated with different resistor values. Note that the TCs of the voltage reference and the resistors are the primary contributors to error.

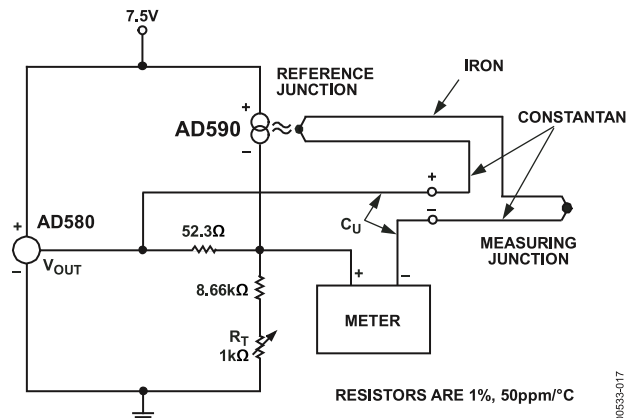


Figure 20. Cold Junction Compensation Circuit for Type J Thermocouple

Transil™, transient voltage surge suppressor (TVS)

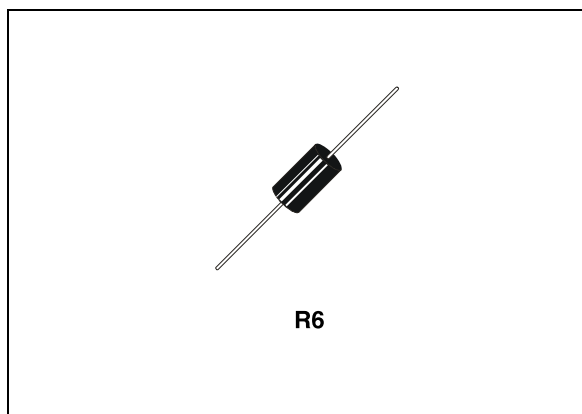
Datasheet – production data

Features

- Peak pulse power: 5000 W (10/1000 μ s)
- Stand-off voltage range from 10 V to 180 V
- Unidirectional and bidirectional types
- Low clamping factor
- Fast response time
- UL497B, file number: QVGQ2.E136224

Description

Transil diodes provide high overvoltage protection by clamping action. Their instantaneous response to transient overvoltages makes them particularly suited to protect voltage sensitive devices such as MOS Technology and low voltage supplied ICs.



TM:Transil is a trademarks of STMicroelectronics.

1 Characteristics

Table 1. Absolute maximum ratings ($T_{amb} = 25\text{ }^{\circ}\text{C}$)

Symbol	Parameter	Value	Unit
P_{PP}	Peak pulse power dissipation ⁽¹⁾	$T_{j\text{ initial}} = T_{amb}$ 5000	W
P	Power dissipation on infinite heatsink	$T_{amb} = 75\text{ }^{\circ}\text{C}$ 6.5	W
I_{FSM}	Non repetitive surge peak forward current for unidirectional types	$t_p = 10\text{ ms}$ $T_{j\text{ initial}} = T_{amb}$ 500	A
T_{stg}	Storage temperature range	-65 to + 175	$^{\circ}\text{C}$
T_j	Maximum operating junction temperature	175	$^{\circ}\text{C}$
T_L	Maximum lead temperature for soldering during 10 s at 5 mm from case.	260	$^{\circ}\text{C}$

1. For a surge greater than the maximum values, the diode will fail in short-circuit.

Table 2. Thermal resistances

Symbol	Parameter	Value	Unit
$R_{th(j-l)}$	Junction to leads	15	$^{\circ}\text{C}/\text{W}$
$R_{th(j-a)}$	Junction to ambient on printed circuit. $L_{lead} = 10\text{ mm}$	65	

Figure 1. Electrical characteristics - definitions

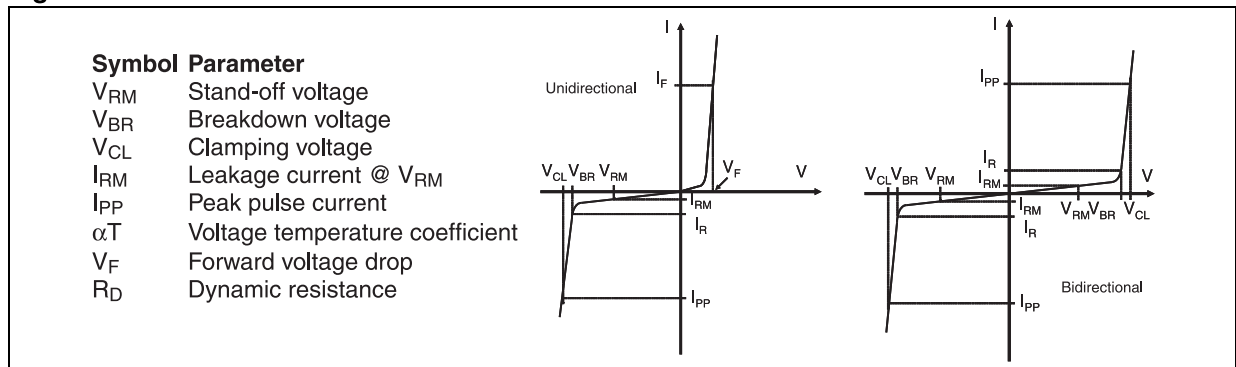


Figure 2. Pulse definition for electrical characteristics

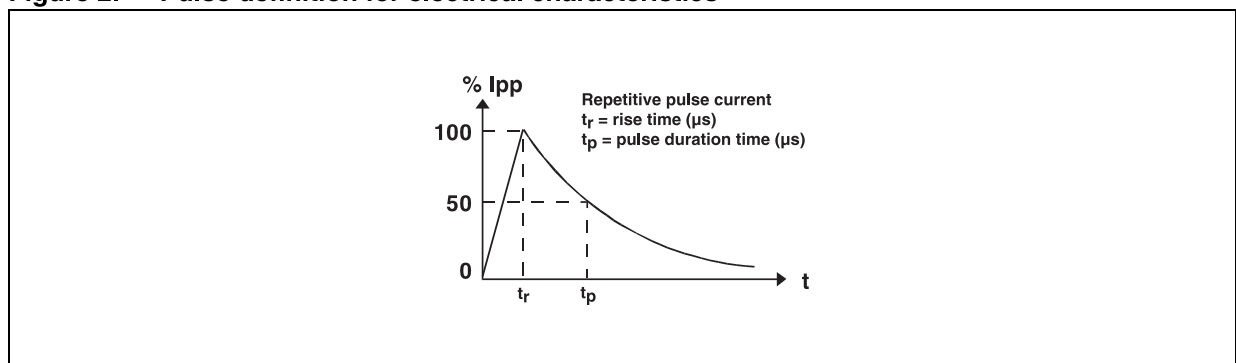


Table 3. Electrical characteristics - values ($T_{amb} = 25\text{ °C}$)

Order code		I_{RM} @ V_{RM}		V_{BR} @ $I_R^{(1)}$		V_{CL} @ I_{PP} 10/1000 μs		V_{CL} @ I_{PP} 8/20 μs		$\alpha T^{(2)}$	$C^{(3)}$
		max		min		max		max		max	typ
Unidirectional	Bidirectional	μA	V	V	mA	V	A	V	A	$10^{-4}/\text{°C}$	pF
BZW50-10	BZW50-10B	5	10	11.1	1	18.8	266	23.4	2564	7.8	24000
BZW50-12	BZW50-12B	5	12	13.3	1	22	227	28	2143	8.4	18500
BZW50-15	BZW50-15B	5	15	16.6	1	26.9	186	35	1714	8.8	13500
BZW50-18	BZW50-18B	5	18	20	1	32.2	155	41.5	1446	9.2	11500
BZW50-22	BZW50-22B	5	22	24.4	1	39.4	127	51	1177	9.6	8500
BZW50-27	BZW50-27B	5	27	30	1	48.3	103	62	968	9.8	7000
BZW50-33	BZW50-33B	5	33	36.6	1	59	85	76	789	10	5750
BZW50-39	BZW50-39B	5	39	43.3	1	69.4	72	90	667	10.1	4800
BZW50-47	BZW50-47B	5	47	52	1	83.2	60.1	108	556	10.3	4100
BZW50-56	BZW50-56B	5	56	62.2	1	99.6	50	129	465	10.4	3400
BZW50-68	BZW50-68B	5	68	75.6	1	121	41	157	382	10.5	3000
BZW50-82	BZW50-82B	5	82	91	1	145	34	189	317	10.6	2600
BZW50-100	BZW50-100B	5	100	111	1	179	28	228	263	10.7	2300
BZW50-120	BZW50-120B	5	120	133	1	215	23	274	219	10.8	1900
BZW50-150	BZW50-150B	5	150	166	1	269	19	343	175	10.8	1700
BZW50-180	BZW50-180B	5	180	200	1	322	16	410	146	10.8	1500

1. Pulse test: $t_p < 50\text{ ms}$.

2. $\Delta V_{BR} = \alpha T \cdot (T_{amb} - 25) \cdot V_{BR}(25\text{°C})$.

3. $V_R = 0\text{ V}$, $F = 1\text{ MHz}$. For bidirectional types, capacitance value is divided by 2.

NTE577 Silicon Diode General Purpose, Fast Recovery Switch

Electrical Characteristics:

Maximum Peak Reverse Voltage, P_{RV}	1000V
Maximum Average Rectified Current ($T_A = +50^\circ$, half-wave, resistive load 60Hz), I_O	5A
Maximum Forward Peak Surge Current (8.3ms superimposed), I_{FSM}	200A
Maximum Reverse Current ($T_A = +25^\circ\text{C}$, $P_{RV} = 1000\text{V}$), I_R	10 μ A
Maximum Forward Voltage ($T_A = +25^\circ\text{C}$, $I_F = 5\text{A}$), V_{FM}	1.7V
Maximum Reverse Recovery Time, t_{rr}	70ns
Operating Temperature Range, T_{opr}	-65° to +150°C
Storage Temperature Range, T_{stg}	-65° to +150°C

